## The Open University of Sri Lanka0.32Faculty of Engineering Technology0.32Department of Electrical & Computer Engineering



Study Programme Name of the Examination **Course Code and Title** Academic Year Date Time Duration Bachelor of Technology Honours in Engineering
Final Examination
EEX7434 Digital Signal Processing
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1400-1700 hrs
3 hours

## **General Instructions**

- 1. Read all instructions carefully before answering the questions.
- 2. This question paper consists of Seven (7) questions in Five (5) pages.
- 3. Answer Any Five (5) questions. All questions carry equal marks.
- 4. Answer for each question should commence from a new page.
- 5. Relevant additional information is provided at the end of the paper.
- 6. This is a Closed Book Test (CBT).
- 7. Answers should be in clear hand writing.
- 8. Do not use red colour pens.

- Q1. (a) Consider the signal  $x(t) = 4\cos(100\pi t)\cos(200\pi t)$ . If the signal x(t) is sampled with a sampling frequency  $f_s$  Hz, express the condition that should be satisfied by the sampling frequency  $f_s$  in order to ensure the perfect recovery of x(t) from the sampled signal. (Hint: You may use the trigonometric identity  $2\cos(\theta)\cos(\phi) = \cos(\theta + \phi) + \cos(\theta \phi)$ .) [4]
  - (b) Consider the two signals  $x_1(t) = 2\cos(40\pi t)$  and  $x_2(t) = 2\cos(70\pi t)$ .
    - i. If the signals  $x_1(t)$  and  $x_2(t)$  are sampled with  $100\pi$  rad/s, find the digital angular frequencies (measured in rad/sample) of the corresponding discrete-time signals  $x_1[n]$  and  $x_2[n]$ . [4]
    - ii. Consider the signal x[n] defined by  $x[n] = x_1[n] + x_2[n]$ . Sketch the frequency spectrum  $X(e^{j\omega})$  of x[n] in the range  $-3\pi \le \omega \le 3\pi$ , where  $\omega$  is the digital angular frequency. [8]
    - iii. Assume that the signal x[n] is applied to an ideal reconstruction filter of which the output is  $\hat{x}(t) = A_1 \cos(\Omega_1 t) + A_2 \cos(\Omega_2 t)$ , where  $A_1, A_2, \Omega_1$ , and  $\Omega_2$  are constants. What are the values of analog angular frequencies  $\Omega_1$  and  $\Omega_2$ ? [4]
- Q2. (a) Briefly explain the linearity and the time-invariance properties of a system. [3]
  - (b) State whether the systems described by the following difference equations are linear and timeinvariant. [4]
    - i. y[n] = 3x[n+1] + 2x[n] + x[n-1]
    - ii. y[n] = nx[n] + x[n-4]
  - (c) State whether systems described by the following difference equations are causal. [3]
    - i. y[n] = 3x[n] + 2x[n-1] + x[n-2]
    - ii. y[n] = x[3n] + x[n-4]
    - iii.  $y[n] = x[n^2] + x[n-2]$
  - (d) Consider an LTI system with the impulse response h[n] and the input signal x[n] as shown in Figure Q2(d). Find the output signal y[n] for  $\forall n \in \mathbb{Z}$ . [10]



Figure Q2(d): The impulse response and the input signal for Q2(d).

[1]

[8]

- Q3. (a) Consider the signal  $x[n] = a^n u[n]$ , where a is a constant.
  - i. Find the *z*-transform, with the region of convergence (ROC), of the signal *x*[*n*] using the first principles.
     [4]
  - ii. Sketch the ROC in the complex plane.
  - (b) Using the answer obtained for Q3(a) and the relevant properties of the *z*-transform, find the *z*-transform (with the ROC) of the signal
     [5]

$$y[n] = \left(\frac{1}{2}\right)^n u[n] + \left(\frac{-1}{3}\right)^n u[n].$$

(c) Consider an LTI system having the transfer function

$$H(z) = \frac{1}{\left(1 - \frac{1}{2}z^{-1}\right)\left(1 - \frac{1}{4}z^{-1}\right)}, \qquad |z| > \frac{1}{2}.$$

Using the answer obtained for Q3(a) and the relevant properties of the z-transform, find the output signal  $y\{n\}$  of the system if the input signal is u[n]. [10]

- Q4. Consider the signal  $x[n] = a^n u[n]$ , where *a* is a real-valued constant and u[n] is the discrete-time unitstep function.
  - (a) For the case |a| < 1, derive the discrete-time Fourier transform (DTFT)  $X(e^{j\omega})$  of x[n] using the first principles. [5]
  - (b) Using the answer obtained for Q4(a), determine  $|X(e^{j\omega})|$  and  $\angle X(e^{j\omega})$ . [5]
  - (c) For the case  $|a| \ge 1$ , can the DTFT  $X(e^{j\omega})$  of x[n] be derived? Justify your answer. [4]
  - (d) Another signal y[n] is defined as  $y[n] = a^{|n|}$ . Express the DTFT  $Y(e^{j\omega})$  in terms of  $X(e^{j\omega})$ . [6]
- **Q5**. (a) Compute the discrete Fourier transform (DFT) of each of the following discrete-time signals having a finite-length, i.e.,  $0 \le n \le (N-1)$ , where N is even:
  - i.  $x[n] = \delta[n n_0]$ , where  $n_0$  is a constant, and  $0 \le n_0 \le (N 1)$ ,
  - ii.  $x[n] = a^n$ ,

iii.  $y[n] = e^{j\omega_0 n}$ , where  $\omega_0$  is a constant, and  $-\pi \le \omega_0 < \pi$ . [12]

(b) A discrete-time system is characterized by the transfer function

$$H(z) = \frac{z^4}{4z^4 - 2z^3 + 3z^2 - z + 2}$$

Determine the stability of the system using the Jury-Marden criterion.

[3]

- Q6. (a) The coefficients of an *N*th order (length (N + 1), where *N* is even) FIR filter H(z) is denoted as  $h[n], -N/2 \le n \le N/2$ . Express the condition that should be satisfied by the coefficients h[n] of the FIR filter in order to have a zero-phase response. [3]
  - (b) The ideal frequency response of a zero-phase lowpass filter H(z) is specified as

$$H_{I}(e^{j\omega}) = \begin{cases} 1, & \text{for } |\omega| \le \omega_{c} \\ 0, & \text{for } \omega_{c} < |\omega| \le \pi, \end{cases}$$

where  $0 < \omega_c < \pi$  is the cutoff frequency of the lowpass filter. Derive a closed-form expression for the infinite-extent ideal impulse response  $h_I[n]$  using the first principles. [7]

(c) A finite-extent impulse response h[n] of length (N+1) can be obtained by multiplying  $h_I[n]$  with an appropriate window function w[n] of length (N+1). Obtain the finite-extent impulse response h[n] (or the coefficients) of a 6th order zero-phase FIR lowpass filter having a cutoff frequency  $0.3\pi$  rad/sample. Use the Hamming window as the window function. The Hamming window of length (N+1) is defined as

$$w[n] = \begin{cases} 0.54 + 0.46 \cos\left(\frac{2\pi n}{N}\right), & \text{for } |n| \le \frac{N}{2} \\ 0, & \text{otherwise.} \end{cases}$$

Provide your answer in a table having columns for  $h_1(n)$ , w(n) and h(n) for required *n*. [7]

- (d) Note that a zero-phase FIR filter is noncausal. How is such a filter converted to a causal filter without changing the magnitude response of the filter. [3]
- Q7. A continuous-time elliptic lowpass filter  $H_c(s)$  having the transfer function

$$H_c(s) = \frac{0.07(s^2 + 2.58)}{(s + 0.38)(s^2 + 0.31s + 0.51)}$$

is employed to design a discrete-time IIR lowpass filter H(z) using the bilinear transform method. The passband edge  $\Omega_p$  of  $H_c(s)$  is 0.71 rad/s, and the sampling frequency is  $10\pi$  rad/s. Furthermore, H(z) is realized as a *cascade structure* of a first-order section and a second-order section.

- (a) Derive the transfer function H(z). [10]
- (b) Determine the passband edge  $\omega_p$  of H(z).
- (c) Draw the realization of H(z) as a cascade structure. Note that the first-order and the second-order sections should be realized using the *direct form II* realizations. [7]

## **Useful Formulae**

• Discrete-Time Fourier Transform (DTFT) of x[n]

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

+ Inverse Discrete-Time Fourier Transform (IDTFT) of  $X(e^{j\omega})$ 

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} \mathrm{d}\omega$$

• *z*-Transform of *x*[*n*]

$$X(z) = \sum_{n = -\infty}^{\infty} x[n] z^{-n}$$

• *N*-point DFT of *x*[*n*]

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi kn}{N}}, \ 0 \le k \le N-1$$

• N-point Inverse DFT (IDFT) of X[k]

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j \frac{2\pi k n}{N}}, \ 0 \le n \le N-1$$

Bilinear transform

$$s = \frac{2}{T} \left( \frac{z-1}{z+1} \right),$$

where T is the sampling period.

• Jury-Marden Criterion: In a stable, linear decrete-time system, all poles are located inside the unit circle centered at the origin.

## END OF THE PAPER